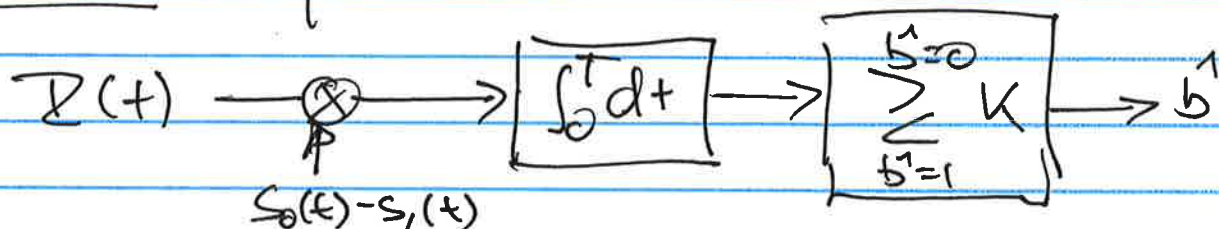


11/28/18

# Signal Space Interpretation of Opt. Receiver

Recall: opt. receiver



For  $\pi_0 = \pi_1 = \frac{1}{2}$  :  $K = \frac{1}{2}(E_0 - E_1)$

$$P_e = Q\left(\sqrt{\frac{E_0 + E_1 - 2\sqrt{E_0 E_1}}{2N_0}}\right)$$

$$P_e = Q\left(\frac{\|s_0(t) - s_1(t)\|}{\sqrt{2N_0}}\right)$$

where  $\|s_0(t) - s_1(t)\|^2 = \int_0^T (s_0(t) - s_1(t))^2 dt$

Given signal set:  $s_0(t), s_1(t)$

Q: Can we find  $\phi_0(t), \phi_1(t)$  such that:

1)  $\langle \phi_0(t), \phi_1(t) \rangle = 0$

2)  $\|\phi_0(t)\| = \|\phi_1(t)\| = 1$

3)  $s_0(t) = A_{00} \phi_0(t) + A_{01} \phi_1(t)$

$s_1(t) = A_{10} \phi_0(t) + A_{11} \phi_1(t)$

A: Yes, it is always possible to find such  $\phi_0(t), \phi_1(t)$

"Proof": The Gram-Schmidt constructs  $\phi_0(t), \phi_1(t)$  from  $s_0(t), s_1(t)$ :

$$a) \phi_0(t) = \frac{s_0(t)}{\|s_0(t)\|} \leftarrow \text{for unit energy}$$

$$b) a) \gamma_1(t) = s_1(t) - \langle s_1(t), \phi_0(t) \rangle \cdot \phi_0(t)$$

$$b) \phi_1(t) = \frac{\gamma_1(t)}{\|\gamma_1(t)\|}$$

Step 1. a) guarantees:  $\gamma_1(t)$  is orth. to  $\phi_0(t)$

$$\langle \gamma_1(t), \phi_0(t) \rangle = \langle \underbrace{s_1(t) - \langle s_1(t), \phi_0(t) \rangle \cdot \phi_0(t)}_{\gamma_1(t)}, \phi_0(t) \rangle$$

$$= \langle s_1(t), \phi_0(t) \rangle - \langle \underbrace{\langle s_1(t), \phi_0(t) \rangle}_{\text{const.}} \cdot \underbrace{\phi_0(t)}_{\|\phi_0(t)\|=1}, \phi_0(t) \rangle$$

$$= \langle s_1(t), \phi_0(t) \rangle - \langle s_1(t), \phi_0(t) \rangle \cdot \underbrace{\langle \phi_0(t), \phi_0(t) \rangle}_{=1}$$

$$= 0$$



To find coefficient  $A_{ij}$  in

$$S_0(t) = A_{00} \cdot \phi_0(t) + A_{01} \phi_1(t)$$

$$S_1(t) = A_{10} \phi_0(t) + A_{11} \phi_1(t)$$

$$A_{00} = \langle S_0(t), \phi_0(t) \rangle$$

$$A_{01} = \langle S_0(t), \phi_1(t) \rangle$$

$$A_{11} = \langle S_1(t), \phi_1(t) \rangle$$

$$\langle S_0(t), \phi_0(t) \rangle = \langle A_{00} \phi_0(t) + A_{01} \phi_1(t), \phi_0(t) \rangle$$

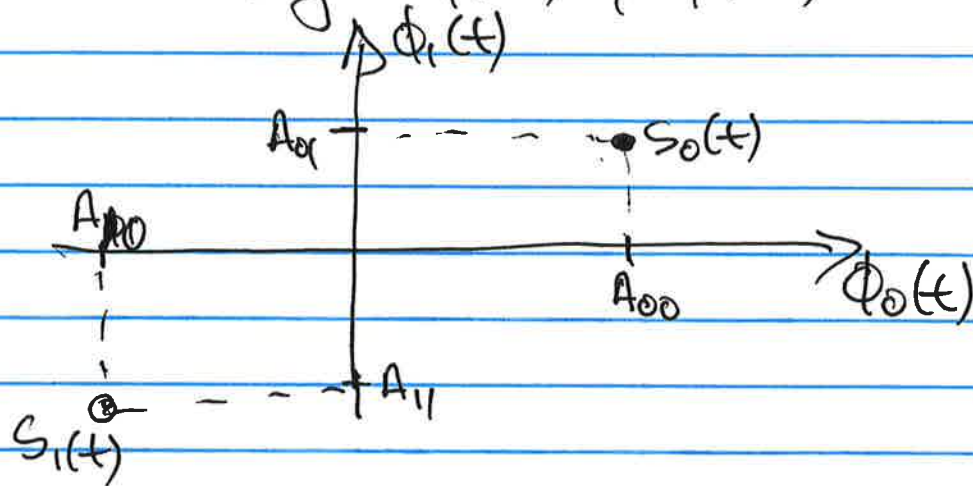
$$= A_{00} \cdot \underbrace{\langle \phi_0(t), \phi_0(t) \rangle}_{=1} + A_{01} \cdot \underbrace{\langle \phi_0(t), \phi_1(t) \rangle}_{=0}$$

$$\Rightarrow \begin{pmatrix} S_0(t) \\ S_1(t) \end{pmatrix}$$

$$S_0(t) = (A_{00}, A_{01}) \cdot \begin{pmatrix} \phi_0(t) \\ \phi_1(t) \end{pmatrix}$$

$$S_1(t) = (A_{10}, A_{11}) \cdot \begin{pmatrix} \phi_0(t) \\ \phi_1(t) \end{pmatrix}$$

This allows us to interpret signals  $s_0(t)$ ,  $s_1(t)$  as "vectors" in a space spanned by  $\phi_0(t)$ ,  $\phi_1(t)$



Note: (Parseval's relationship)

$$\begin{aligned}
 E_i = \|s_i(t)\|^2 &= \int_0^T s_i^2(t) dt = \int_0^T (A_{0i}\phi_0(t) + A_{1i}\phi_1(t))^2 dt \\
 &= \int_0^T A_{0i}^2 \phi_0^2(t) dt + 2A_{0i}A_{1i} \int_0^T \phi_0(t)\phi_1(t) dt + A_{1i}^2 \int_0^T \phi_1^2(t) dt \\
 &= A_{0i}^2 + A_{1i}^2 = \| \underline{s}_i \|^2 \\
 \underline{s}_i &= \begin{pmatrix} A_{0i} \\ A_{1i} \end{pmatrix}
 \end{aligned}$$

Similarly,  $\langle s_0(t), s_1(t) \rangle = \langle \underline{s}_0, \underline{s}_1 \rangle$



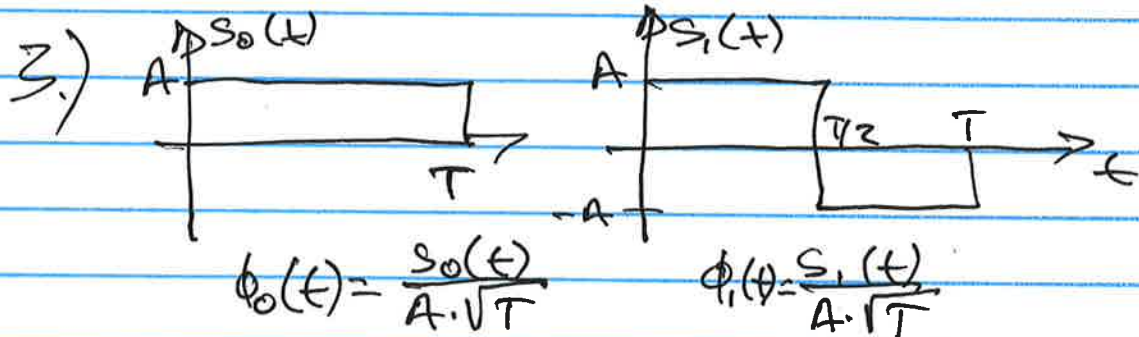
# Examples:

$\cos(\cdot)$   
and  
 $\sin(\cdot)$   
are orth.

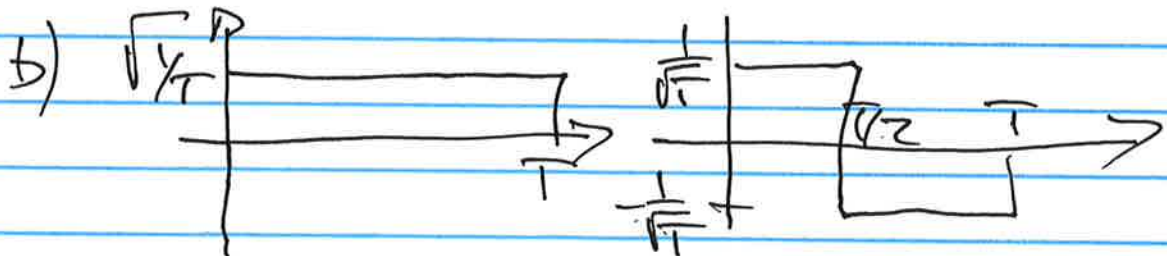
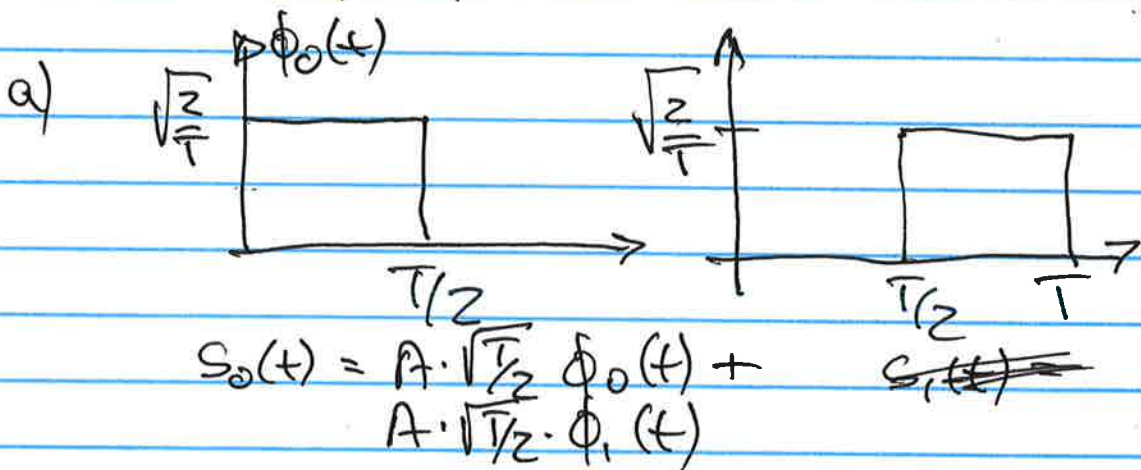
$$1. \left. \begin{aligned} s_0(t) &= \sqrt{E} \cdot \sqrt{\frac{2}{T}} \cos(2\pi f_c t) \\ s_1(t) &= \sqrt{E} \cdot \sqrt{\frac{2}{T}} \sin(2\pi f_c t) \end{aligned} \right\} 0 \leq t \leq T$$

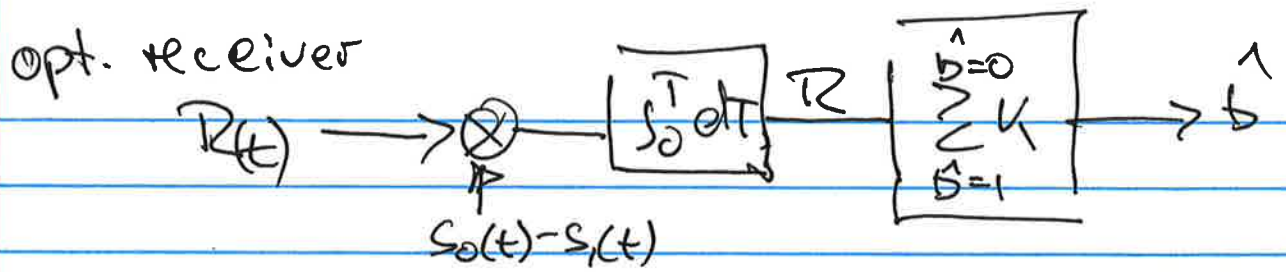
$$2. \left. \begin{aligned} s_0(t) &= \sqrt{E} \cdot \sqrt{\frac{2}{T}} \cos(2\pi f_0 t) \\ s_1(t) &= \sqrt{E} \cdot \sqrt{\frac{2}{T}} \cos(2\pi f_1 t) \end{aligned} \right\} \begin{array}{l} \phi_0(t) \\ \phi_1(t) \end{array}$$

$f_0 \cdot T$  and  $f_1 \cdot T$  are distinct integers



For 3.) possible orthonormal bases:





$$R = \langle R(t), s_0(t) - s_1(t) \rangle$$

$$= \langle R(t), s_0(t) \rangle - \langle R(t), s_1(t) \rangle$$

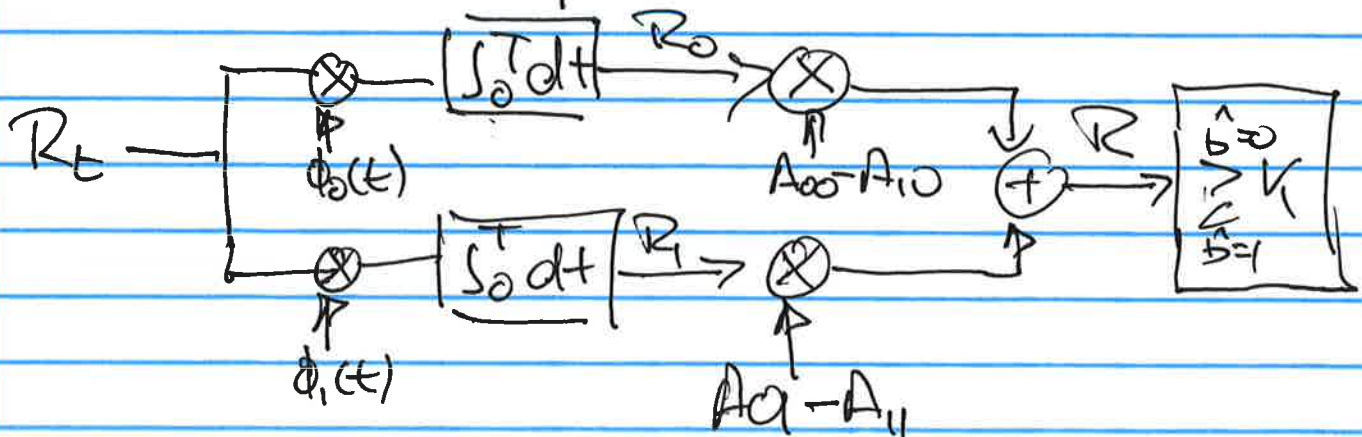
$$= \langle R(t), A_{00}\phi_0(t) + A_{01}\phi_1(t) \rangle - \langle R(t), A_{10}\phi_0(t) + A_{11}\phi_1(t) \rangle$$

$$= A_{00} \cdot \langle R(t), \phi_0(t) \rangle + A_{01} \langle R(t), \phi_1(t) \rangle - A_{10} \langle R(t), \phi_0(t) \rangle - A_{11} \langle R(t), \phi_1(t) \rangle$$

$$= (A_{00} - A_{10}) \cdot \underbrace{\langle R(t), \phi_0(t) \rangle}_{R_0} + (A_{01} - A_{11}) \underbrace{\langle R(t), \phi_1(t) \rangle}_{R_1}$$

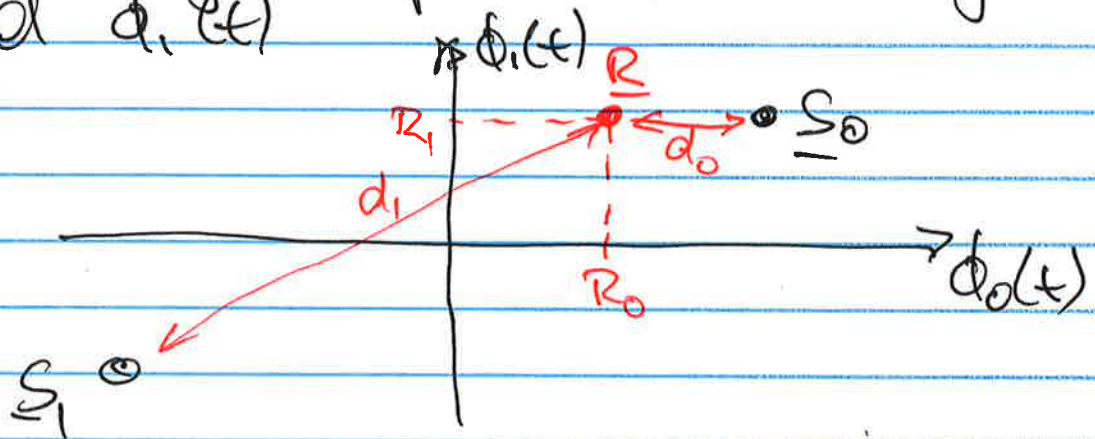
$$= \begin{pmatrix} A_{00} - A_{10} \\ A_{01} - A_{11} \end{pmatrix}^T \cdot \underbrace{\begin{pmatrix} R_0 \\ R_1 \end{pmatrix}}_{\underline{R}} = \langle \underline{s_0} - \underline{s_1}, \underline{R} \rangle$$

$\Rightarrow$  alternative opt. receiver





Frontend projects the received signal into signal space spanned by  $\phi_0(t)$  and  $\phi_1(t)$



For  $\pi_0 = \pi_1 = \frac{1}{2}$ , opt. receiver decides in favor of signal that is closest to  $\underline{R}$ :

$$\Leftrightarrow \|\underline{R} - \underline{s}_0\|^2 \quad \begin{matrix} \sum_{b^1=1}^1 \\ \sum_{b^2=0}^1 \\ \sum_{b^3=1}^1 \\ \sum_{b^4=0}^1 \end{matrix} \quad d_1 \quad \|\underline{R} - \underline{s}_1\|^2$$

$$\Leftrightarrow \|\underline{R}\|^2 - 2\langle \underline{R}, \underline{s}_0 \rangle + \|\underline{s}_0\|^2 \stackrel{\sum_{b^1=1}^1}{\leq} \|\underline{R}\|^2 - 2\langle \underline{R}, \underline{s}_1 \rangle + \|\underline{s}_1\|^2$$

$$\Leftrightarrow \|\underline{s}_0\|^2 - \|\underline{s}_1\|^2 \stackrel{\sum_{b^1=1}^1}{\leq} \sum_{b^1=0}^1 2\langle \underline{R}, \underline{s}_0 \rangle - 2\langle \underline{R}, \underline{s}_1 \rangle$$

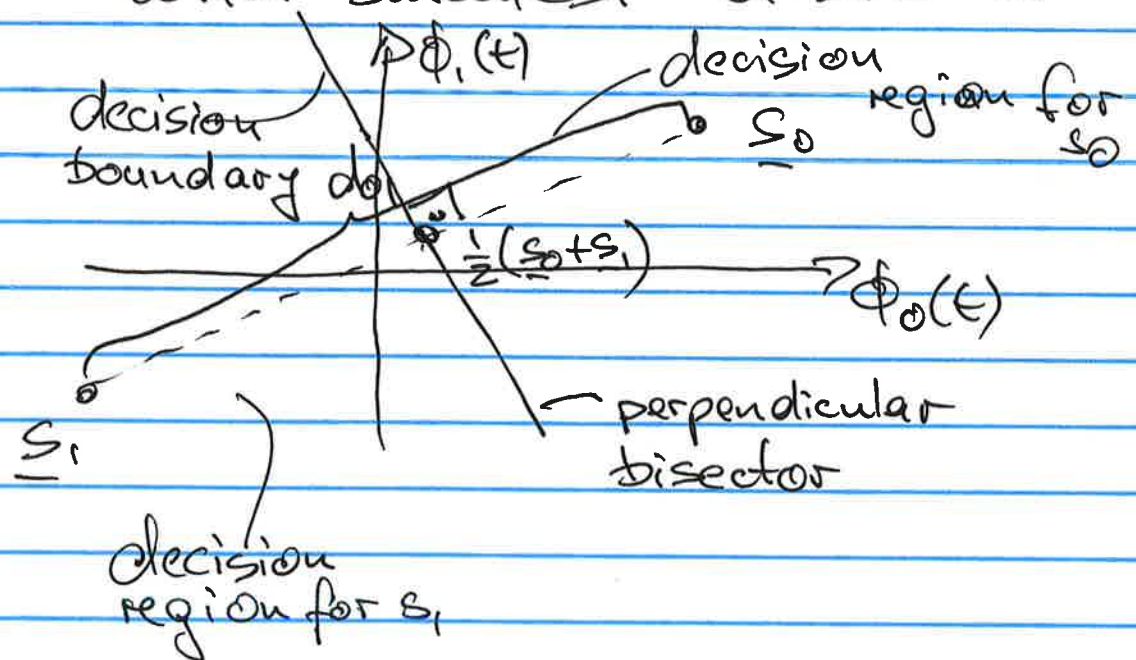
$$\frac{\|\underline{s}_0\|^2 - \|\underline{s}_1\|^2}{2} \stackrel{\sum_{b^1=1}^1}{\leq} \sum_{b^1=0}^1 \langle \underline{R}, \underline{s}_0 - \underline{s}_1 \rangle$$

OPTIMUM RECEIVER!

$$K = \frac{\|\underline{s}_0\|^2 - \|\underline{s}_1\|^2}{2} \stackrel{\sum_{b^1=1}^1}{\leq} \sum_{b^1=0}^1 \langle \underline{R}_t, \underline{s}_d(t) - \underline{s}_1(t) \rangle$$

## Interpretation: of opt receiver

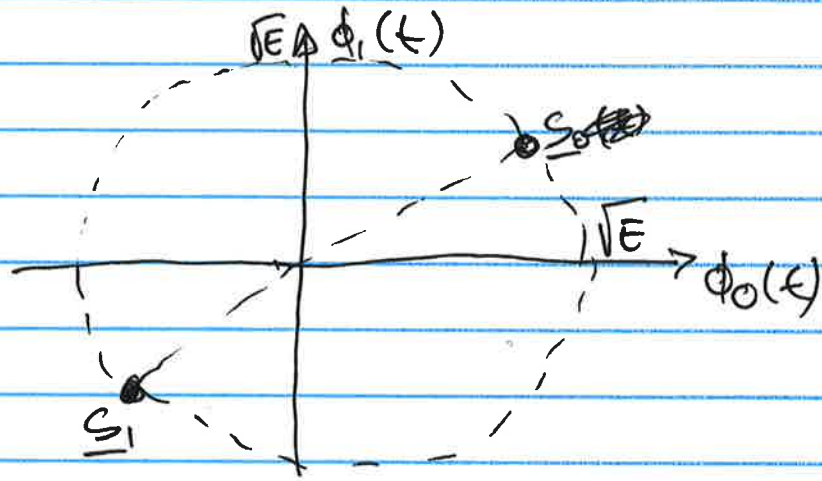
1. Frontend: projects the received signal into the signal space that contains  $s_0(t), s_1(t)$  (spanned by  $\phi_0(t), \phi_1(t)$ )
2. Decision: is based on distances  $d_i = \|R - s_i\|$  between received signal and signals  $s_0(t), s_1(t)$ . Decision is in favor of signal with smallest distance



$$\begin{aligned} \text{Finally, } P_e &= Q\left(\frac{\|s_0 - s_1\|}{\sqrt{2N_0}}\right) \\ &= Q\left(\frac{d_{01}}{\sqrt{2N_0}}\right) \end{aligned}$$



Q: If signals are limited to energy  $E$ , what is the best (min  $P_e$ ) signal set  $s_0(t), s_1(t)$ ?



A: Signals have to be negatives of each other and lie on circle with radius  $\sqrt{E}$

$\Rightarrow$  this is the BPSK signal set.