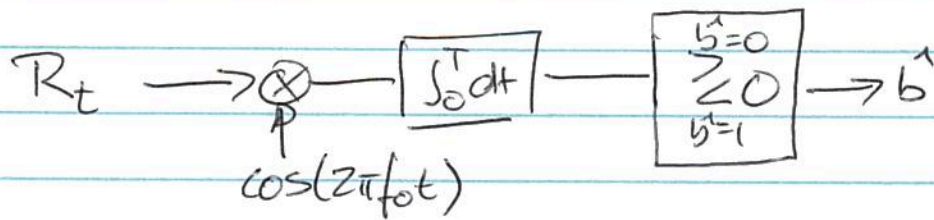


HW 11 solution

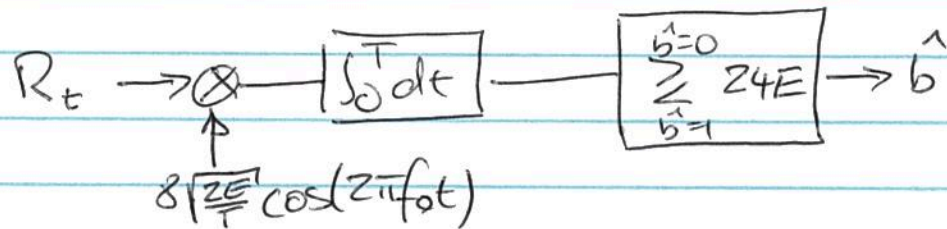
1. a)



$$b) E_0 = E_1 = -\gamma_{01} = 25E$$

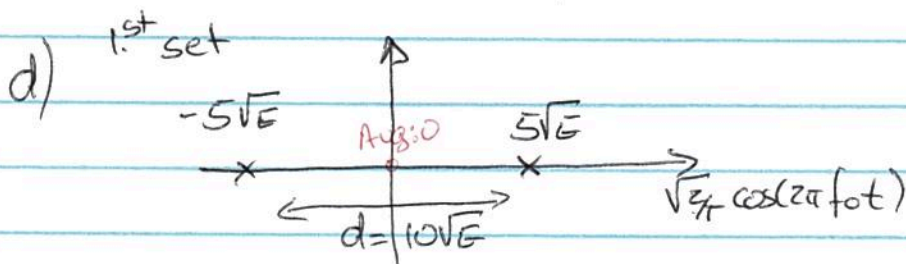
$$\Rightarrow P_e = Q\left(\sqrt{\frac{100E}{2N_0}}\right) = Q\left(\sqrt{\frac{50E}{N_0}}\right)$$

c) opt. receiver

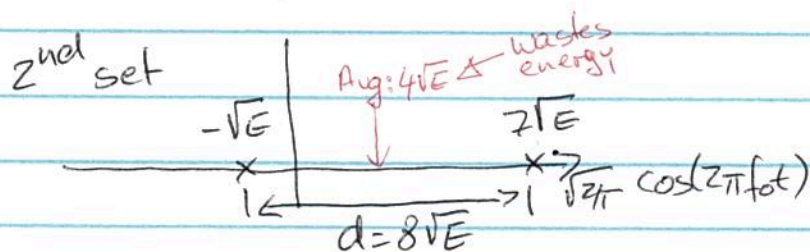


$$E_0 = 49E \quad E_1 = 1 \quad \gamma_{01} = -7E$$

$$P_e = Q\left(\sqrt{\frac{64E}{2N_0}}\right) = Q\left(\sqrt{\frac{32E}{N_0}}\right)$$



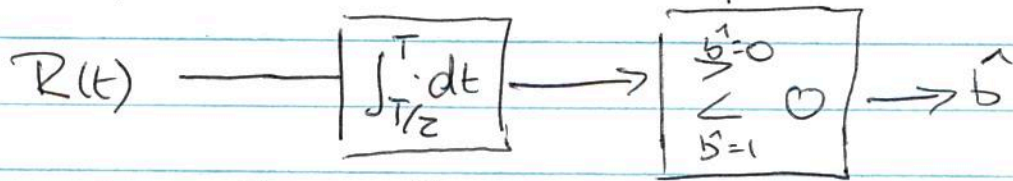
$$e) E_b = \frac{1}{2}(E_0 + E_1) = 25E$$



$$E_b = \frac{1}{2}(E_0 + E_1) = 25E$$

1st set is better; for same transmitted energy (25E),
1st set has better distance \Rightarrow better P_e

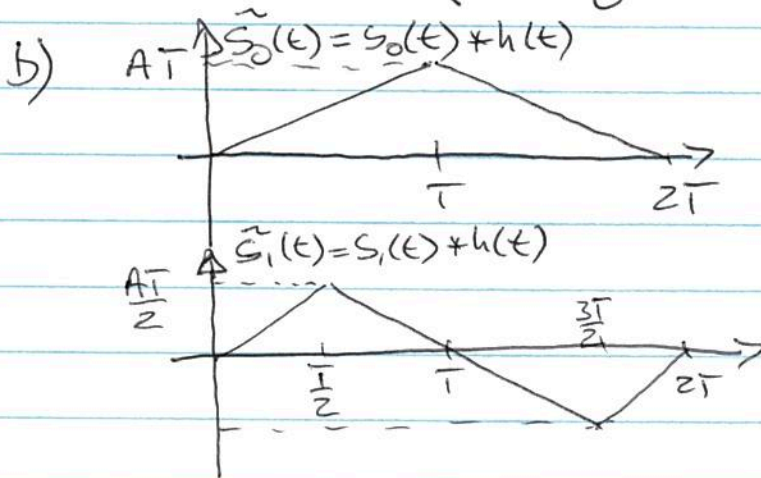
2. a) Opt. receiver can be simplified to: (why?)



$$E_0 = E_1 = A^2 T$$

$$\Gamma_{01} = 0$$

$$\Rightarrow P_e = Q\left(\sqrt{\frac{2A^2 T}{2N_0}}\right) = Q\left(A\sqrt{\frac{T}{N_0}}\right)$$



c) The receiver in (a) is not optimal for the signals $\tilde{s}_0(t)$ and $\tilde{s}_1(t)$.

$$P_e = Q\left(\frac{\langle \tilde{s}_0(t) - \tilde{s}_1(t), g(t) \rangle}{\sqrt{2N_0 \|g(t)\|^2}}\right) \quad \text{with: } g(t) = \begin{cases} 1, & \frac{T}{2} \leq t \leq T \\ 0, & \text{else} \end{cases}$$

$$\langle \tilde{s}_0(t) - \tilde{s}_1(t), g(t) \rangle = \int_{T/2}^T \tilde{s}_0(t) - \tilde{s}_1(t) dt$$

$$= \int_{T/2}^T A \cdot t - (AT - At) dt$$

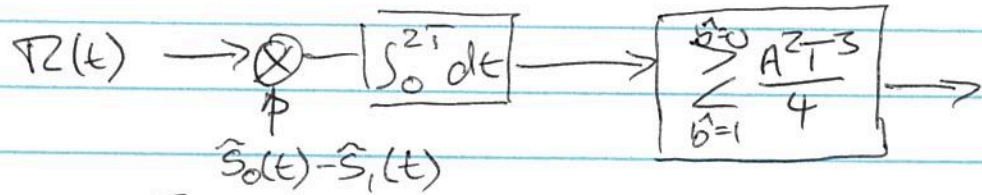
$$= \int_{T/2}^T 2At - AT dt$$

$$= \frac{2At^2}{2} - ATt \Big|_{T/2}^T = \frac{AT^2}{4}$$

$$\|g(t)\|^2 = \int_{T/2}^T 1^2 dt = \frac{T}{2}$$

$$\Rightarrow P_e = Q\left(\frac{AT^2/4}{\sqrt{2N_0 \cdot T/2}}\right) = Q\left(\frac{AT}{4} \cdot \sqrt{\frac{T}{N_0}}\right) = Q\left(AT \cdot \sqrt{\frac{T}{16 \cdot N_0}}\right)$$

d) Opt. receiver correlates with $\tilde{S}_0(t) - \tilde{S}_1(t)$! Note, this requires that $h(t)$ is known.



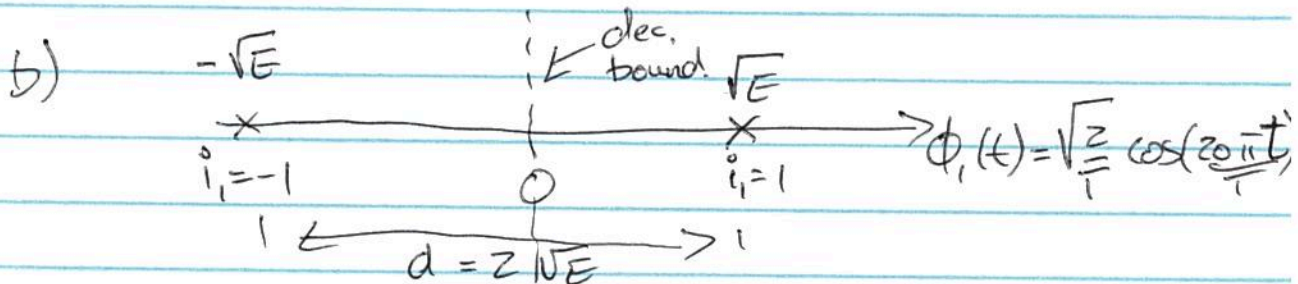
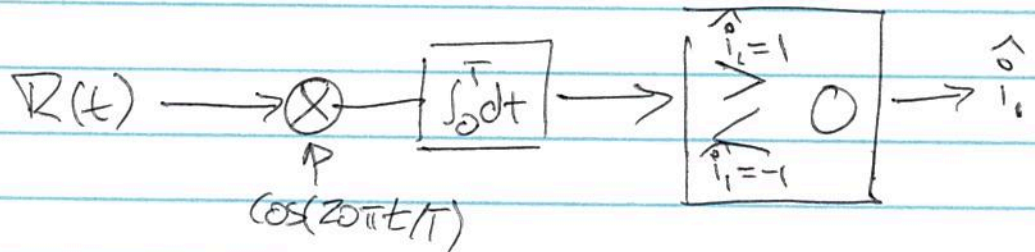
$$\tilde{E}_0 = 2 \cdot \int_0^T (At)^2 dt = \frac{2}{3} A^2 T^3 \quad \tilde{\Gamma}_0 = 0$$

$$\tilde{E}_1 = 4 \cdot \int_0^{T/2} (At)^2 dt = \frac{1}{6} A^2 T^3 \quad \tilde{K} = \frac{1}{2} (\tilde{E}_0 - \tilde{E}_1) = \frac{A^2 T^3}{4}$$

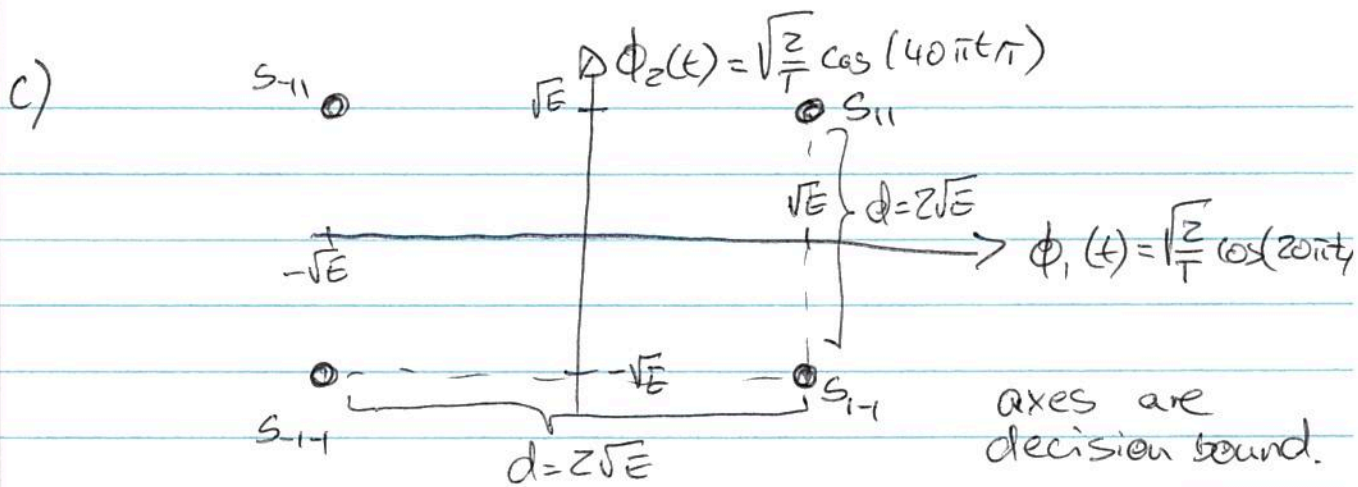
$$P_e = Q\left(\sqrt{\frac{\tilde{E}_0 + \tilde{E}_1}{2N_0}}\right) = Q\left(AT \cdot \sqrt{\frac{5-T}{12 \cdot N_0}}\right)$$

This is much (namely $\frac{20}{3}$ times better SNR) than (c)

3. a)

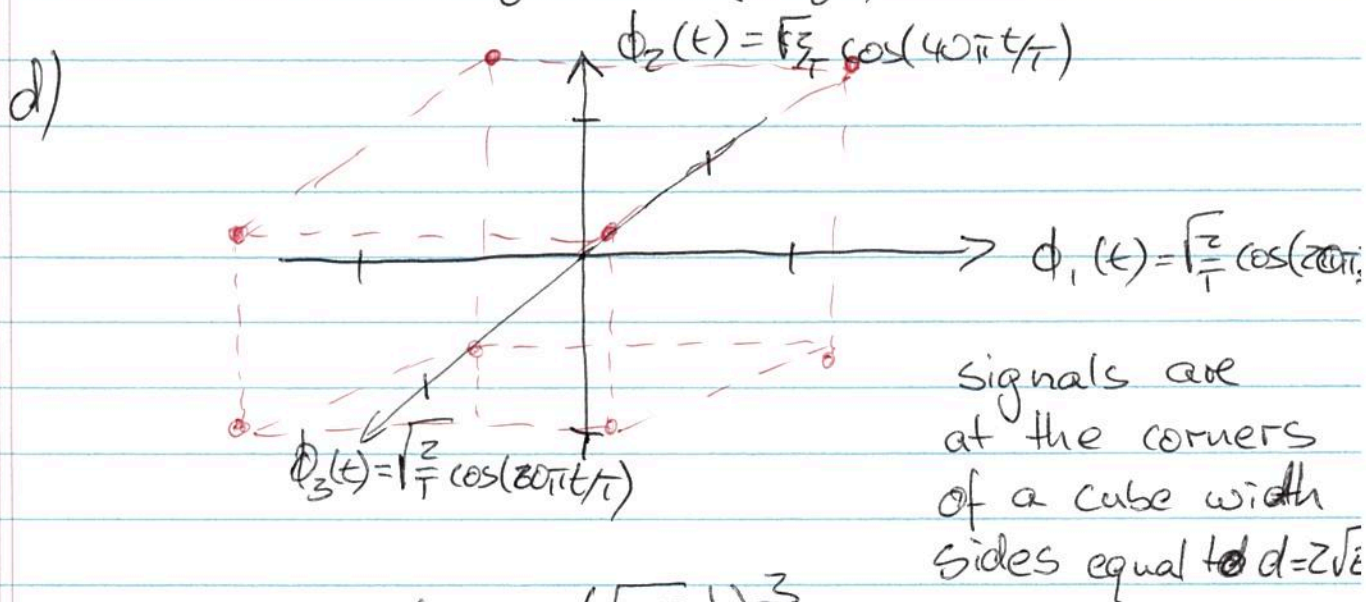


$$P_e = Q\left(\frac{d}{\sqrt{2N_0}}\right) = Q\left(\frac{2\sqrt{E}}{\sqrt{2N_0}}\right) = Q\left(\sqrt{\frac{2E}{N_0}}\right)$$



$$P_e = 1 - (1 - Q(\sqrt{\frac{2E}{N_0}}))^2$$

$$= 2Q(\sqrt{\frac{2E}{N_0}}) - Q^2(\sqrt{\frac{2E}{N_0}})$$



$$P_e = 1 - (1 - Q(\sqrt{\frac{2E}{N_0}}))^3$$

$$= 3Q(\sqrt{\frac{2E}{N_0}}) - 3Q^2(\sqrt{\frac{2E}{N_0}}) + Q^3(\sqrt{\frac{2E}{N_0}})$$

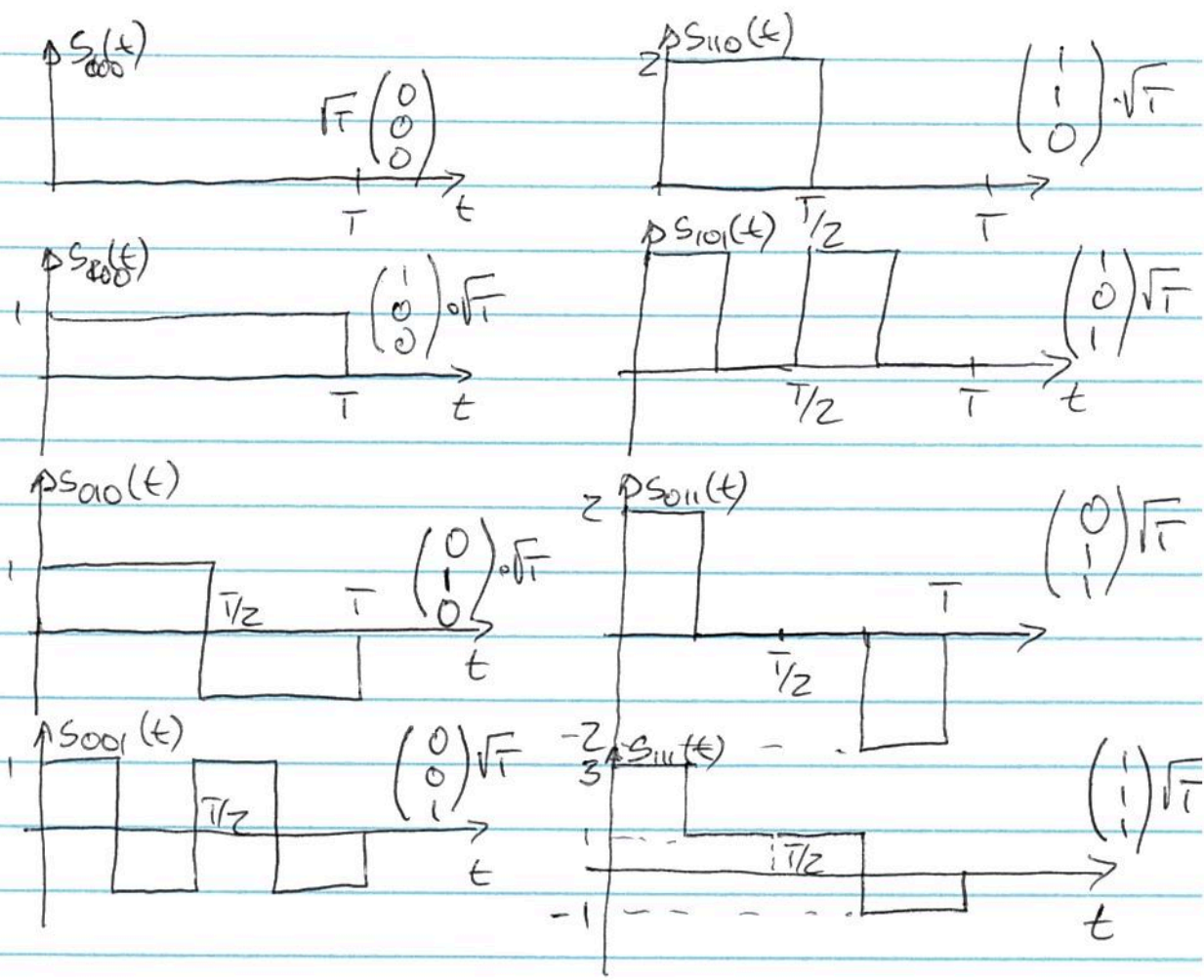
e) General expression for error probability

$$P_e = 1 - (1 - Q(\sqrt{\frac{2E}{N_0}}))^N$$

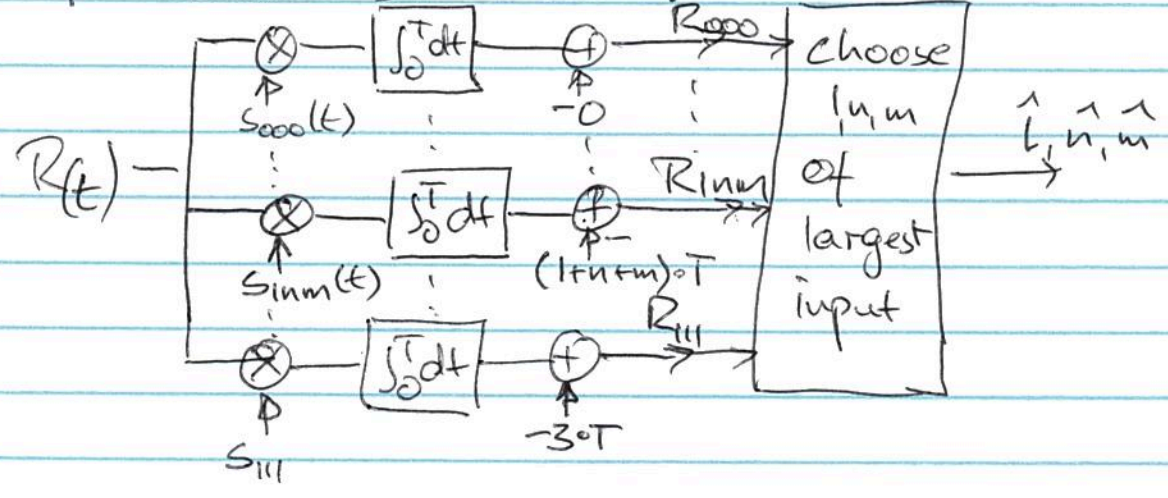
$$= \sum_{n=1}^N (-1)^n \binom{N}{n} Q^n(\sqrt{\frac{2E}{N_0}})$$

signals are the corners of N -dimensional signal hypercube.

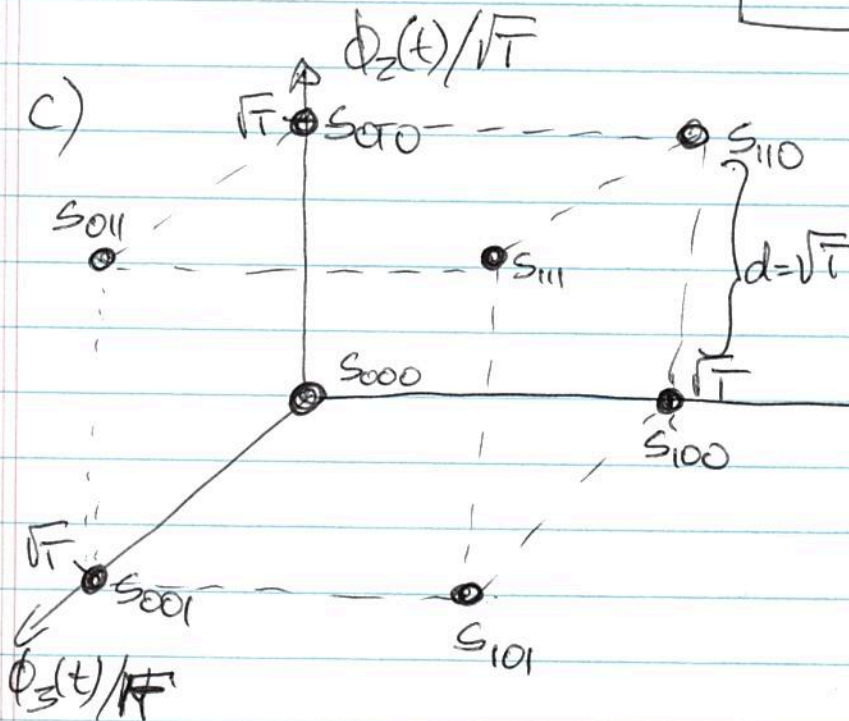
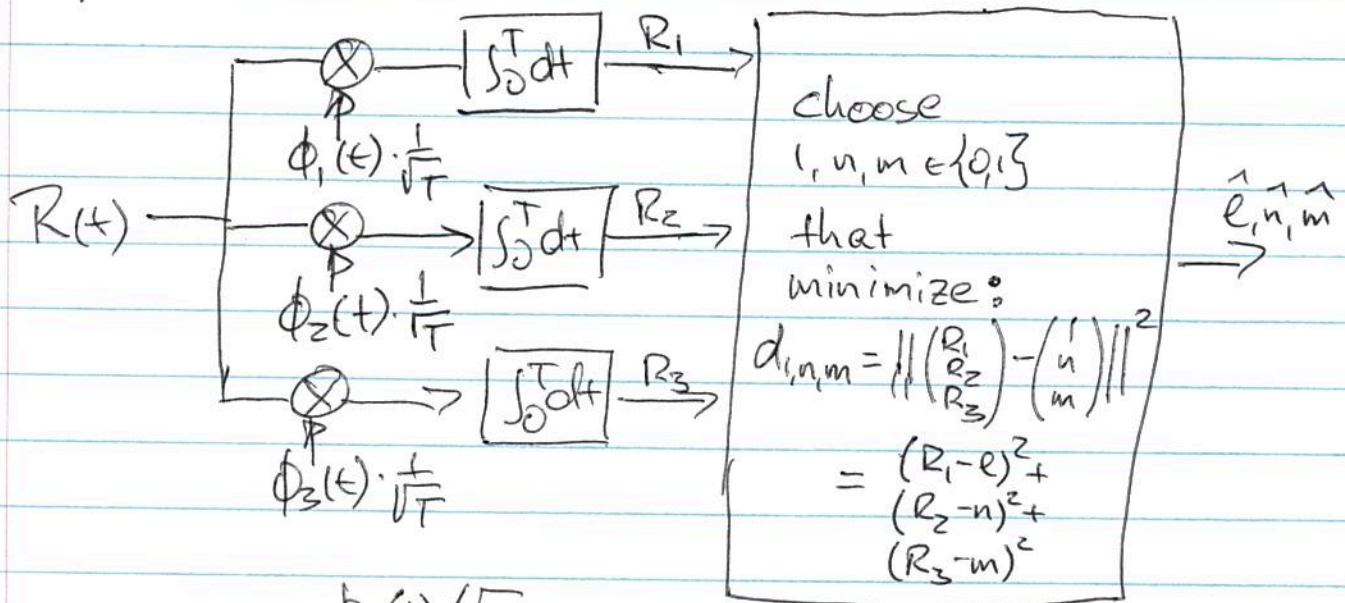
4a)



b) Opt. receiver structure 1: 8 correlators



opt. receiver structure 2: 3 correlators



- decision boundaries are the three planes that cut the cube into eight equal pieces
- decision boundaries are perpendicular to sides of cube
- sides of cube are length $d = \sqrt{T}$

$$d) P_e = 1 - \left(1 - Q\left(\frac{\sqrt{T}}{\sqrt{2N_0}}\right)\right)^3$$

$$= 3Q\left(\frac{\sqrt{T}}{\sqrt{2N_0}}\right) - 3Q^2\left(\frac{\sqrt{T}}{\sqrt{2N_0}}\right) + Q^3\left(\frac{\sqrt{T}}{\sqrt{2N_0}}\right)$$

$$e) s_{l,n,m}(t) = \sqrt{2} \cdot l \cdot \cos(20\pi t/T) + \sqrt{2} \cdot n \cdot \cos(40\pi t/T) + \sqrt{2} \cdot m \cdot \cos(60\pi t/T) \quad l, n, m \in \{0, 1\}$$