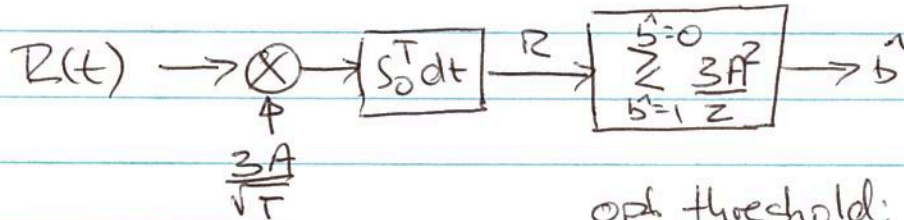


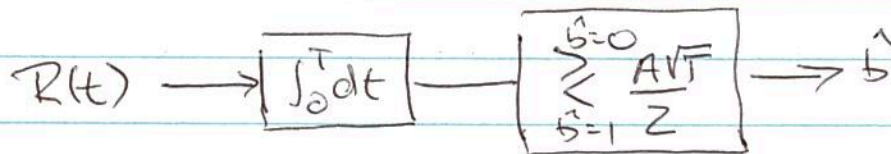
1.

a) opt. receiver



$$\begin{aligned} \text{opt. threshold: } k &= \frac{1}{2} (\bar{E}_0 - \bar{E}_1) \\ &= \frac{1}{2} (4A^2 - A^2) \\ &= \frac{3}{2} A^2 \end{aligned}$$

alternatively, we can absorb the constant in the correlator into the threshold:



b) For the optimum receiver, the prob. of error is given by:

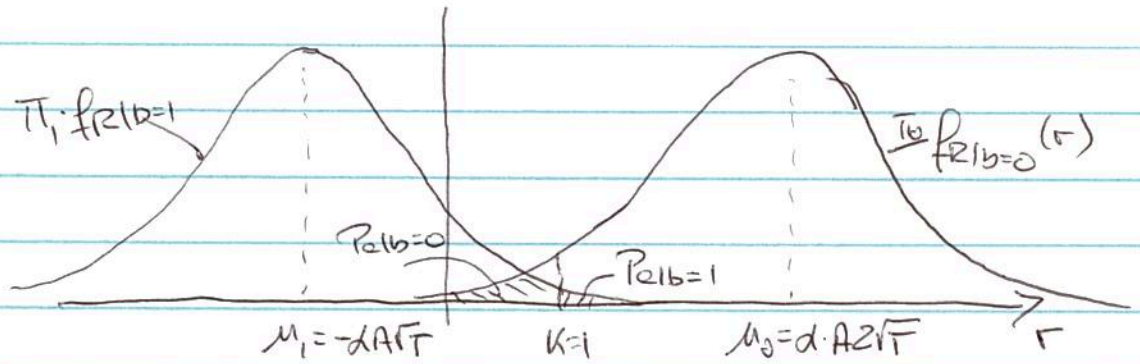
$$\begin{aligned} P_e &= Q\left(\sqrt{\frac{\bar{E}_0 + \bar{E}_1 - 2\sigma_{01}}{2N_0}}\right) \\ &= Q\left(\frac{3A}{\sqrt{2N_0}}\right) \end{aligned}$$

$$\begin{aligned} \bar{E}_0 &= 4A^2 \\ \bar{E}_1 &= A^2 \\ \sigma_{01} &= 2A^2 \end{aligned}$$

c) Conditioned on either  $b=0$  or  $b=1$ ,  $R$  will be Gaussian with the following means and variances:

	$b=0$	$b=1$
Mean: $\mu_0, \mu_1$	$\alpha \cdot 2A \cdot \sqrt{T}$	$-\alpha \cdot A \cdot \sqrt{T}$
Variance: $\sigma^2$	$\alpha^2 \cdot \frac{N_0}{2} \cdot T$	$\alpha^2 \cdot \frac{N_0}{2} \cdot T$

d)



For the fixed threshold ( $k=1$ ), we cannot assume that conditional error probabilities  $P_{e|b=0}$  and  $P_{e|b=1}$  are equal  $\Rightarrow$  compute both

$$P_{e|b=1} = \int_{-\infty}^{\infty} f_{r|b=1}(r|b=1) dr$$

$$= \int_{-\infty}^{\infty} \frac{1}{\sqrt{2\pi}\sigma} \exp\left(-\frac{(r-\mu_1)^2}{2\sigma^2}\right) dr$$

$$z = \frac{r-\mu_1}{\sigma}; \quad = \int_{\frac{1-\mu_1}{\sigma}}^{\infty} \frac{1}{\sqrt{2\pi}} e^{-z^2/2} dz = Q\left(\frac{1-\mu_1}{\sigma}\right) = Q\left(\frac{1+\alpha A\sqrt{T}}{\alpha \cdot \sqrt{N_0/2} \cdot T}\right)$$

$$P_{e|b=0} = \int_{-\infty}^1 \frac{1}{\sqrt{2\pi}\sigma} \exp\left(-\frac{(r-\mu_0)^2}{2\sigma^2}\right) dr$$

$$z = -\frac{r-\mu_0}{\sigma}; \quad \int_{\frac{\mu_0-1}{\sigma}}^{\infty} \frac{1}{\sqrt{2\pi}} e^{-z^2/2} dz = Q\left(\frac{\mu_0-1}{\sigma}\right) = Q\left(\frac{\alpha A\sqrt{T}-1}{\alpha \sqrt{N_0/2} T}\right)$$

$\Rightarrow$  overall  $P_e$ :

$$P_e = \frac{1}{2} \cdot \left( Q\left(\frac{1+\alpha A\sqrt{T}}{\alpha \sqrt{N_0/2} T}\right) + Q\left(\frac{\alpha A\sqrt{T}-1}{\alpha \sqrt{N_0/2} T}\right) \right)$$



e) The gain  $\alpha$  can be used to minimize  $P_e$ .  
 For minimum  $P_e$ , the two conditional error probabilities must be equal. Equivalently, we can use  $\alpha$  to ensure that the fixed threshold  $k=1$  is optimal.

For equally likely bits the optimal threshold is the average of the means:

$$k_{opt} = \frac{1}{2} (\mu_0 + \mu_1) = \frac{1}{2} (\alpha z A \sqrt{T} - \alpha A \sqrt{T})$$

$$= \alpha \cdot \frac{A \sqrt{T}}{2}$$

We want  $k_{opt}$  to equal the fixed  $k=1$ :

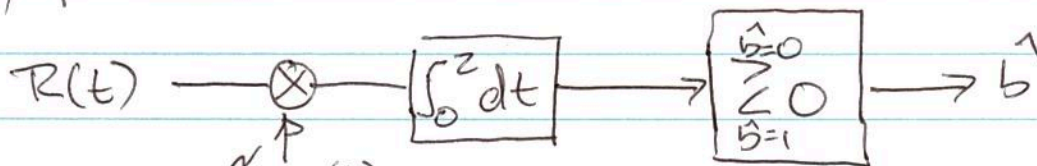
$$\Rightarrow 1 = \alpha \frac{A \sqrt{T}}{2}$$

$$\Rightarrow \alpha = \frac{2}{A \sqrt{T}}$$

Plugging this value of  $\alpha$  into the conditional probabilities of error yields:

$$P_{e|b=0} = P_{e|b=1} = P_e = Q\left(\frac{3A}{\sqrt{2N_0}}\right) \quad (\text{same as (b)} \Rightarrow \text{opt. receiver})$$

2.a) opt. receiver:



$s_0(t)$   
 not needed  
 since threshold  
 is 0.

opt. threshold is 0  
 since  $s_1(t) = -s_0(t)$   
 $\Rightarrow E_0 = E_1$ .

$$b) E_0 = E_1 = \int_0^2 s_0^2(t) dt$$

$$= 4 \cdot \int_0^{1/2} (2t)^2 dt$$

$$= 4 \cdot \frac{4t^3}{3} \Big|_0^{1/2} = \frac{16}{3} \cdot \left(\frac{1}{2}\right)^3 = \frac{2}{3}$$

$$r_{01} = \int_0^2 s_0(t) \cdot s_1(t) dt = - \int_0^2 s_0^2(t) dt = -E_0 = -\frac{2}{3}$$

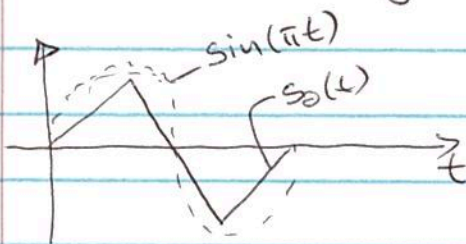
$s_1(t) = -s_0(t)$

$$\Rightarrow P_e = Q\left(\sqrt{\frac{E_0 + E_1 - 2r_{01}}{2N_0}}\right) = Q\left(\sqrt{\frac{4 \cdot 2/3}{2N_0}}\right) = Q\left(\sqrt{\frac{4}{3N_0}}\right)$$

c) For this receiver, the probability of error equals:

$$P_e = Q\left(\frac{\langle s_0(t) - s_1(t), g(t) \rangle}{\sqrt{2N_0 \cdot \|g(t)\|^2}}\right)$$

$$\langle s_0(t) - s_1(t), g(t) \rangle = \int_0^2 2 \cdot s_0(t) \cdot \sin(\pi t) dt$$



$$= 4 \cdot \int_0^{1/2} 2 \cdot 2t \cdot \sin(\pi t) dt$$

$$= 16 \cdot \left( \frac{\sin(\pi t) - \pi t \cos(\pi t)}{\pi^2} \right) \Big|_0^{1/2}$$

$$= \frac{16}{\pi^2}$$

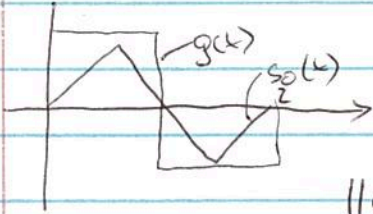
$$\|g(t)\|^2 = \int_0^2 \sin^2(\pi t) dt = \int_0^2 \left(\frac{1}{2} - \frac{1}{2} \cos(2\pi t)\right) dt = 1$$

$$\Rightarrow P_e = Q\left(\frac{16}{\pi^2 \sqrt{2N_0}}\right)$$



d) Proceeding similarly:

$$\langle s_0(t) - s_1(t), g(t) \rangle = \int_0^2 2 \cdot s_0(t) \cdot g(t) dt$$



$$= 4 \cdot \int_0^{1/2} 2 \cdot 2t dt = \frac{16t^2}{2} \Big|_0^{1/2} = 2$$

$$\|g(t)\|^2 = \int_0^2 g^2(t) dt = 2 \cdot \int_0^1 1^2 dt = 2$$

$$\Rightarrow P_e = Q\left(\frac{\langle s_0(t) - s_1(t), g(t) \rangle}{\sqrt{2N_0 \cdot \|g(t)\|^2}}\right) = Q\left(\frac{2}{\sqrt{2N_0 \cdot 2}}\right) = Q\left(\frac{1}{\sqrt{N_0}}\right)$$

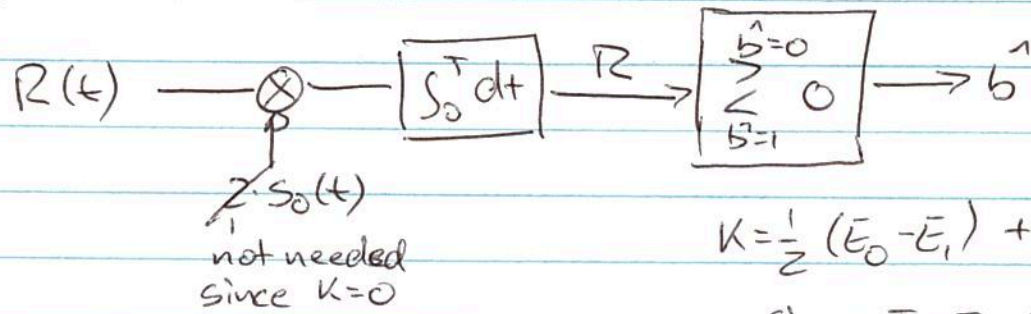
e) Receiver (b) is best (d) is worst.

Consider:

$$\underbrace{\sqrt{\frac{4}{3}} \cdot \frac{1}{\sqrt{N_0}}}_{\approx 1.1547} > \underbrace{\frac{16}{\pi^2 \sqrt{2}} \cdot \frac{1}{\sqrt{N_0}}}_{1.1146} > 1 \cdot \frac{1}{\sqrt{N_0}}$$

- We know, that matching with  $s_0(t) - s_1(t)$ , i.e., triangular signals, is optimum.
- The sin signal in (c) is very similar to triangles  $\Rightarrow$  small loss in performance
- The rectangular signals in (d) are least like the triangles  $\Rightarrow$  worst of the three.

3. a)



$$K = \frac{1}{2} (\bar{E}_0 - \bar{E}_1) + \frac{N_0}{2} \ln \left( \frac{\pi_1}{\pi_0} \right) = 0$$

Since  $\pi_0 = \pi_1$  and  $\bar{E}_0 = \bar{E}_1$

$$b) P_e = Q \left( \frac{\|s_0(t) - s_1(t)\|}{\sqrt{2N_0}} \right)$$

$$\|s_0(t) - s_1(t)\|^2 = \int_0^T (s_0(t) - s_1(t))^2 dt$$

$$= \int_0^T 4 \cdot s_0^2(t) dt$$

$$= 2 \cdot \int_0^{T/2} 4 \cdot (At)^2 dt = \frac{8A^2 t^3}{3} \Big|_0^{T/2} = \frac{AT^3}{3}$$

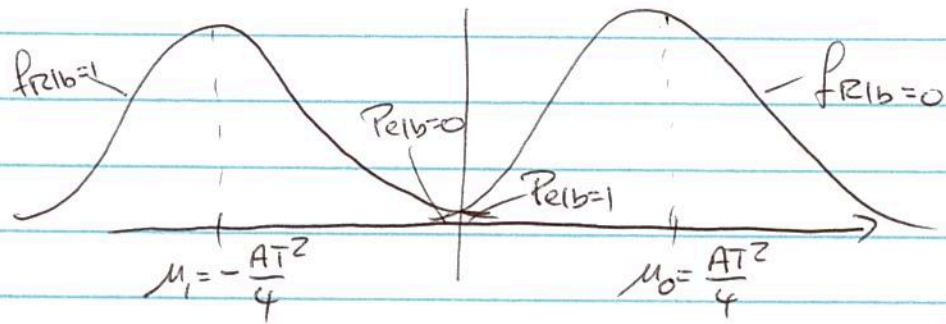
$$\Rightarrow P_e = Q \left( AT \cdot \sqrt{\frac{T}{6N_0}} \right)$$

c) conditional distributions of  $R$  are Gaussian:

	$b=0$	$b=1$
mean:	$\mu_0 = \langle s_0(t), g(t) \rangle$ $= 2 \cdot \int_0^{T/2} At dt$ $= \frac{AT^2}{4}$	$\mu_1 = \langle s_1(t), g(t) \rangle$ $= -\frac{AT^2}{4}$
Variance:	$\sigma^2 = \frac{N_0}{2} \ g(t)\ ^2$ $= \frac{N_0}{2} \cdot T$	$\sigma^2 = \frac{N_0}{2} T$



d)

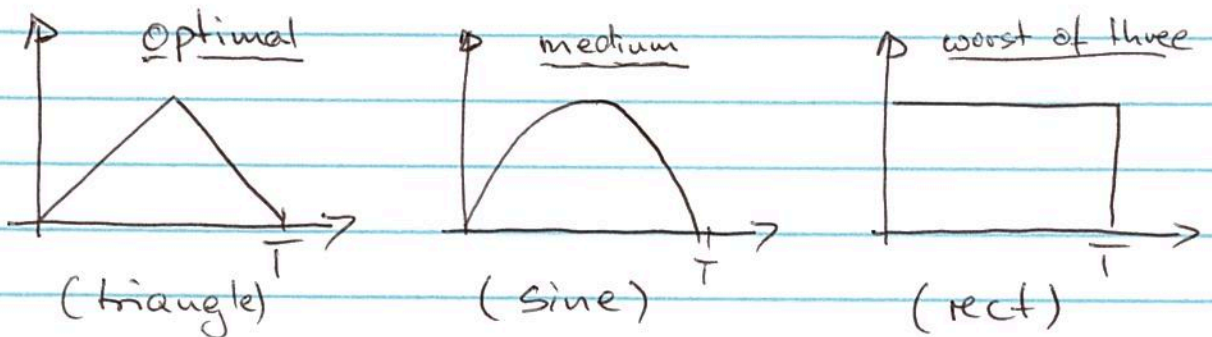


By symmetry,  $P_{e|b=0} = P_{e|b=1} = P_e = Q\left(\frac{\mu_0}{\sigma}\right)$

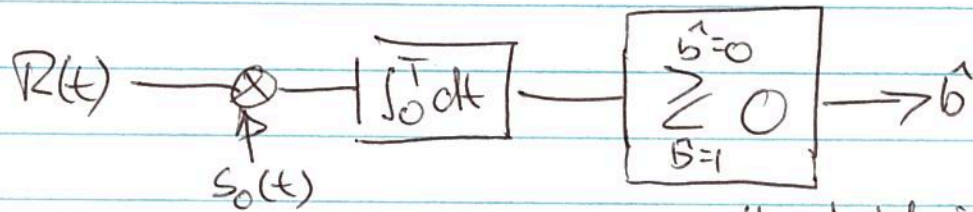
$$= Q\left(\frac{AT^2/4}{\sqrt{N_0/2} \cdot T}\right)$$

$$= Q\left(AT \cdot \sqrt{\frac{T}{8N_0}}\right)$$

e) Sinusoidal pulses would give lower  $P_e$  than the rectangular pulses (but not as good as the optimal triangular pulses). The sine pulse is a better approximation to the triangles than the rectangular pulses:



4. a)



threshold is zero,  
since signals are  
equally likely and  
have equal energy.

$$\begin{aligned}
 b) P_e &= Q\left(\sqrt{\frac{E_0 + E_1 - 2r_{01}}{2N_0}}\right) \\
 &= Q\left(A \cdot \sqrt{\frac{4T}{3 \cdot 2N_0}}\right) \\
 &= Q\left(A \cdot \sqrt{\frac{2T}{3N_0}}\right)
 \end{aligned}$$

$$\begin{aligned}
 E_0 = E_1 = -r_{01} &= \int_0^T s_0^2(t) dt \\
 &= 2 \cdot \int_0^{T/2} \left(\frac{2At}{T}\right)^2 dt \\
 &= 2 \cdot \frac{4A^2}{T^2} \frac{t^3}{3} \Big|_0^{T/2} \\
 &= \frac{A^2 T}{3}
 \end{aligned}$$

c) conditioned on  $b=0$  and  $b=1$ , respectively,  $R$  is Gaussian with following means & variances:

	$b=0$	$b=1$
Mean:	$  \begin{aligned}  \mu_0 &= \int_0^T s_0(t) dt \\  &= 2 \cdot \int_0^{T/2} \left(\frac{2At}{T}\right) dt \\  &= \frac{AT}{2}  \end{aligned}  $	$  \begin{aligned}  \mu_1 &= \int_0^T s_1(t) dt \\  &= -\frac{AT}{2}  \end{aligned}  $
Variance:	$  \begin{aligned}  \sigma^2 &= \frac{N_0}{2} \int_0^T 1 dt \\  &= \frac{N_0}{2} \cdot T  \end{aligned}  $	$  \sigma^2 = \frac{N_0 T}{2}  $



d) By symmetry:  $P_e = P_{e|b=0} = P_{e|b=1}$

$$= Q\left(\frac{\mu_0}{\sigma}\right)$$

$$= Q\left(\frac{AT/2}{\sqrt{N_0/2} \cdot T}\right) = Q\left(A \cdot \sqrt{\frac{T}{2N_0}}\right)$$

e) When the transmitted signal is amplified by  $\alpha$ , then:

- Means in part (c) change to

$$\mu_0 = \alpha \frac{AT}{2} \quad \text{and} \quad \mu_1 = -\alpha \frac{AT}{2}$$

- Variance is unchanged

- Probability of error changes to:

$$P_e = Q\left(\alpha A \cdot \sqrt{\frac{T}{2N_0}}\right)$$

To make this  $P_e$  equal to the optimal  $P_e$  (in part (b)):

$$\alpha \cdot A \cdot \sqrt{\frac{T}{2N_0}} = A \cdot \sqrt{\frac{2T}{3N_0}}$$

$$\Rightarrow \alpha = \sqrt{\frac{4}{3}}$$

f) Part (e) says that with the suboptimum receiver, we must increase transmitted power by  $\alpha^2 = \frac{4}{3}$  to achieve  $P_e$  of optimum receiver. Thus, the suboptimum loses  $\alpha^2 = \frac{4}{3} = 1.25 \text{ dB}$  in relation to the optimum receiver.